

EELE 477
Digital Signal Processing

10b

z-Transforms

Inverse and Deconvolution

- We have seen that two systems in cascade can be combined into a single system by multiplying $H_1(z)$ and $H_2(z)$.
- We can also take a system function $H(z)$ and *factor* it into two or more low-order systems.
- Question: can we *divide* the system output by the system function (“deconvolve”) and recover the input?

$$Y(z) = H_1(z)X(z); \quad Y(z)H_2(z) = X(z) \quad ?$$

$$Y(z) = H_1(z)H_2(z)Y(z) \Rightarrow H_1(z)H_2(z) = 1?$$

Inverse and Deconvolve, cont.

- If we can find $H_2(z)$, it is called the *inverse* of $H_1(z)$.
- NOTE that $H_2(z)$ will not be FIR if $H_1(z)$ is FIR.
- $H_2(z)$ may represent a non-causal and/or unstable system even if $H_1(z)$ is causal and stable.

Relating $H(z)$ and $H(e^{j\omega})$

- NOTE CAREFULLY: z-transform and frequency response formulae are of identical form.

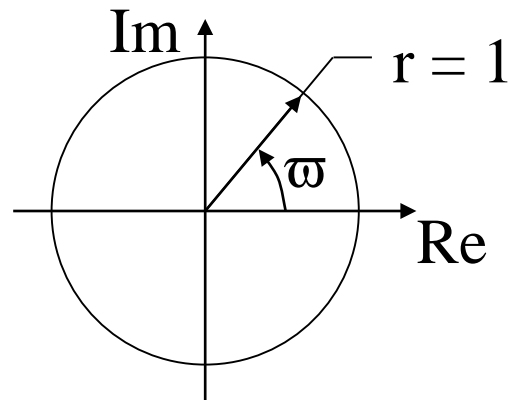
$$H(\hat{\omega}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k} \quad H(z) = \sum_{k=0}^M b_k z^{-k}$$

- If we evaluate $H(z)$ for $z=e^{j\omega}$, it is clear:

$$H(\hat{\omega}) = H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\omega}}$$

Properties of $z=e^{j\omega}$

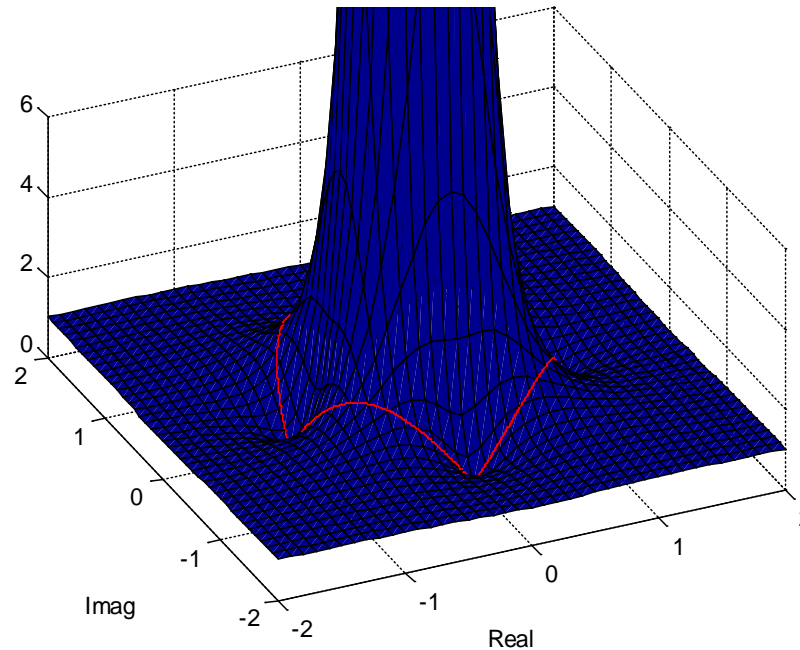
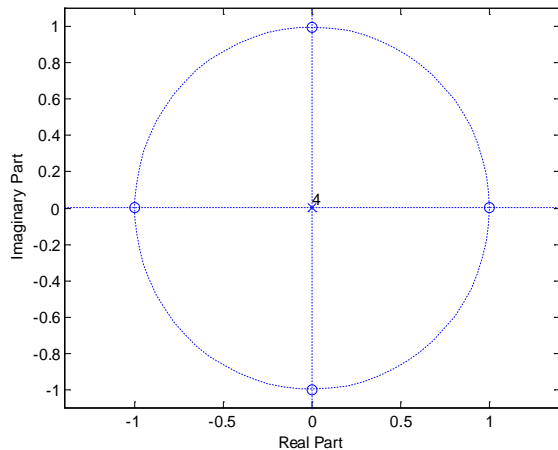
- Observe $z=e^{j\omega}$ for $-\pi < \omega < \pi$: $|z|=1$,
phase= ω
- This defines a *circle* in the z-plane with
radius=1: referred to as the *unit circle*



Visualizing Frequency Response

- We can observe z-transform along the unit circle to reveal the frequency response.

$$H(z) = 1 - z^{-4}$$



Poles and Zeros

- A *pole* in the z -domain is a value of z that “pushes up” the magnitude like a tent pole.
- A *zero* in the z -domain is a value of z that “pins down” the magnitude like a stake or tack.
- The pole and zero locations control the magnitude everywhere, *including along the unit circle*.

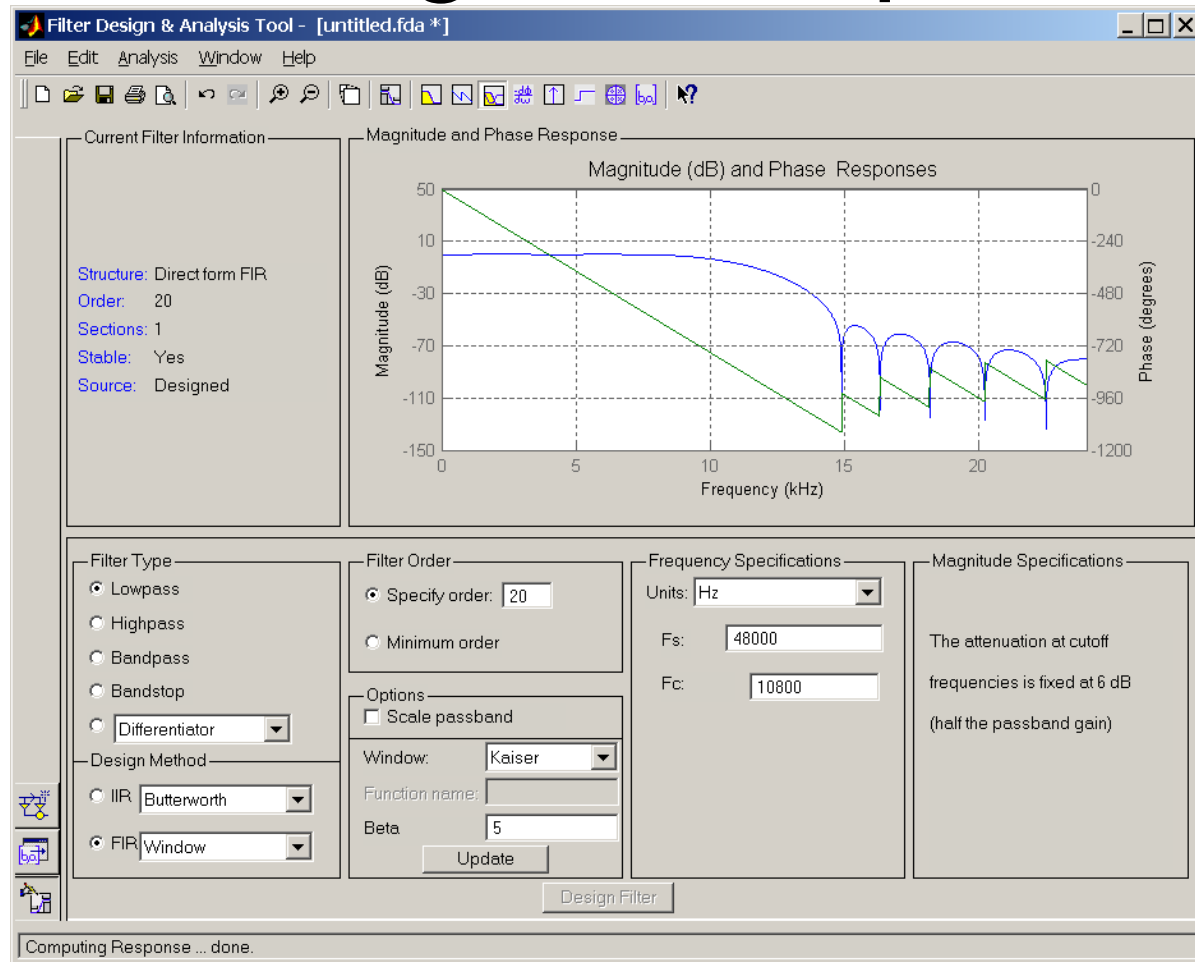
FIR Systems

- FIR systems contain only finite zeros. Poles are located at zero (and perhaps infinity).
- FIR filter design requires a careful choice of zero locations.
- Stop band has zeros on the unit circle.
- Pass band has zeros off the unit circle.

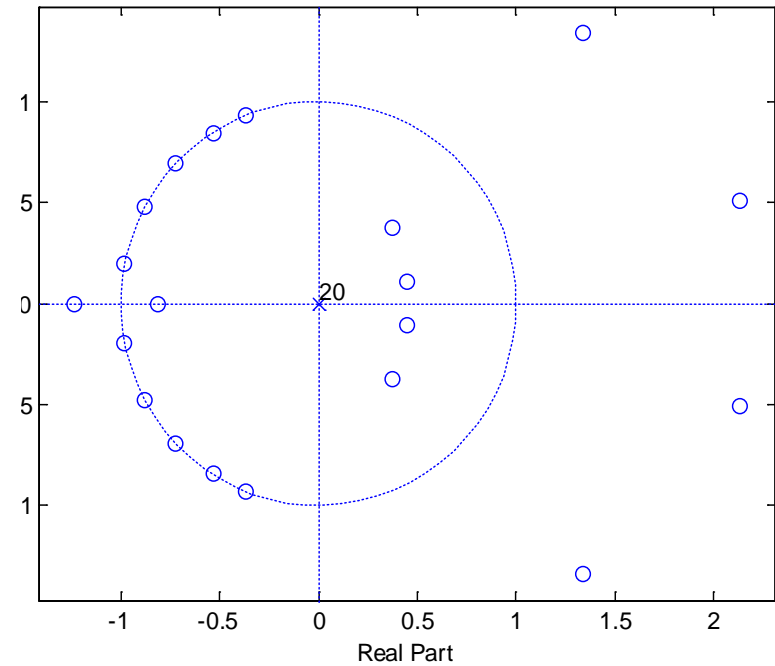
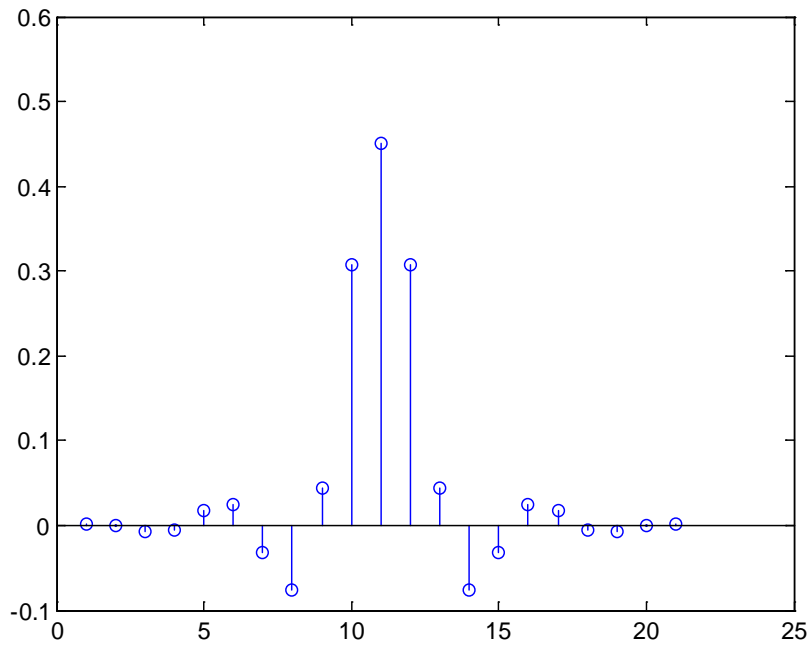
Matlab FIR Filter Design

- Matlab provides several FIR filter design tools, including: `fir1`, `fir2`, and `remez`
- Matlab GUI: Filter Design and Analysis Tool (FDATool)
- Usually specify passband ripple, stopband attenuation, band edges, filter order, and f_s

Design Example



Design Example (cont.)



Symmetry and Linear Phase

- FIR systems with symmetric coefficients ($b_k = b_{M-k}$) have frequency responses with *linear phase*.
- Show this by grouping z-transform terms, for example:

$$\begin{aligned} H(z) &= b_0 + b_1 z^{-1} + b_2 z^{-2} + b_3 z^{-3} + b_2 z^{-4} + b_1 z^{-5} + b_0 z^{-6} \\ &= z^{-3} \left[b_0 (z^3 + z^{-3}) + b_1 (z^2 + z^{-2}) + b_2 (z^1 + z^{-1}) + b_3 \right] \end{aligned}$$

Linear Phase (cont.)

- Now evaluate $H(z)$ on unit circle:

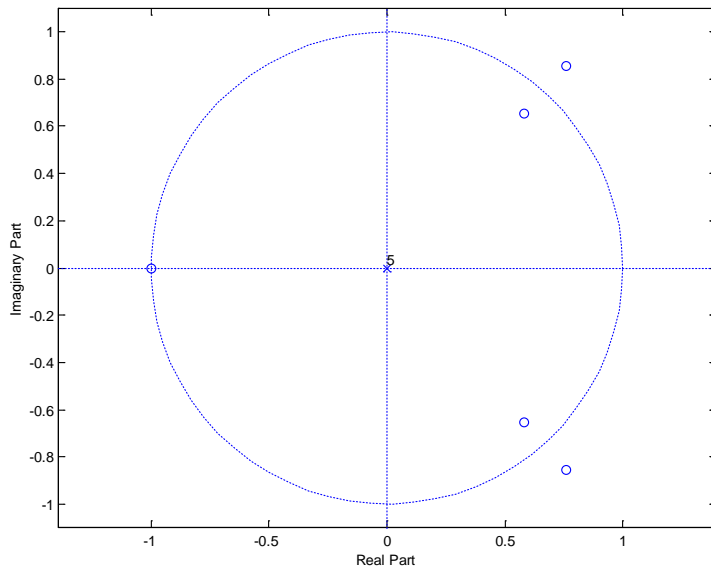
$$H(e^{j\hat{\omega}}) = \underbrace{e^{-j3\hat{\omega}}}_{\substack{\text{linear} \\ \text{phase} \\ \text{term}}} \left[\underbrace{b_0 \left(\underbrace{e^{j3\hat{\omega}} + e^{-j3\hat{\omega}}}_{2\cos(3\hat{\omega})} \right) + b_1 \left(\underbrace{e^{j2\hat{\omega}} + e^{-j2\hat{\omega}}}_{2\cos(2\hat{\omega})} \right) + b_2 \left(\underbrace{e^{j\hat{\omega}} + e^{-j\hat{\omega}}}_{2\cos(\hat{\omega})} \right) + b_3}_{\text{a real function of } \hat{\omega} \text{ (phase=0)}} \right]$$

- Example if M is odd:

$$\begin{aligned} H(z) &= b_0 + b_1 z^{-1} + b_2 z^{-2} + b_2 z^{-3} + b_1 z^{-4} + b_0 z^{-5} \\ &= z^{-2.5} \left[b_0 \left(z^{2.5} + z^{-2.5} \right) + b_1 \left(z^{1.5} + z^{-1.5} \right) + b_2 \left(z^{0.5} + z^{-0.5} \right) \right] \end{aligned}$$

Zero Symmetry

- For an FIR linear phase system (implies coefficient symmetry), the zeros will have a specific pattern. For each z_0 , there will be:



$$\left\{ z_0, z_0^*, \frac{1}{z_0}, \frac{1}{z_0^*} \right\}$$