

EE 477  
Digital Signal Processing

7b

z-Transforms

# Inverse and Deconvolution

- We have seen that two systems in cascade can be combined into a single system by multiplying  $H_1(z)$  and  $H_2(z)$ .
- We can also take a system function  $H(z)$  and *factor* it into two or more low-order systems.
- Question: can we *divide* the system output by the system function (“deconvolve”) and recover the input?

$$Y(z) = H_1(z)X(z); \quad Y(z)H_2(z) = X(z) \quad ?$$

$$Y(z) = H_1(z)H_2(z)Y(z) \Rightarrow H_1(z)H_2(z) = 1?$$

# Inverse and Deconvolve, cont.

- If we can find  $H_2(z)$ , it is called the *inverse* of  $H_1(z)$ .
- NOTE that  $H_2(z)$  will not be FIR if  $H_1(z)$  is FIR.
- $H_2(z)$  may represent a non-causal and/or unstable system even if  $H_1(z)$  is causal and stable.

# Relating $H(z)$ and $H(e^{j\omega})$

- NOTE CAREFULLY: z-transform and frequency response formulae are of identical form.

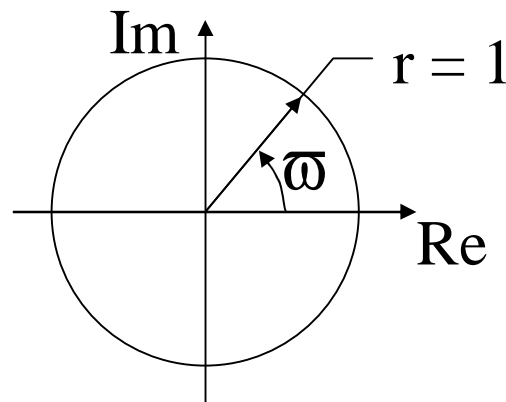
$$H(\hat{\omega}) = \sum_{k=0}^M b_k e^{-j\hat{\omega}k} \quad H(z) = \sum_{k=0}^M b_k z^{-k}$$

- If we evaluate  $H(z)$  for  $z=e^{j\omega}$ , it is clear:

$$H(\hat{\omega}) = H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\omega}}$$

# Properties of $z=e^{j\omega}$

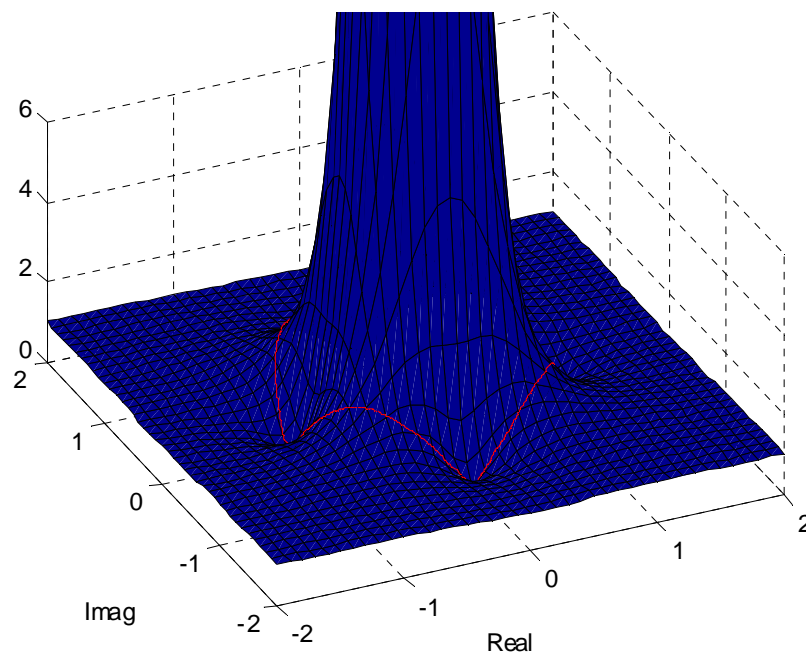
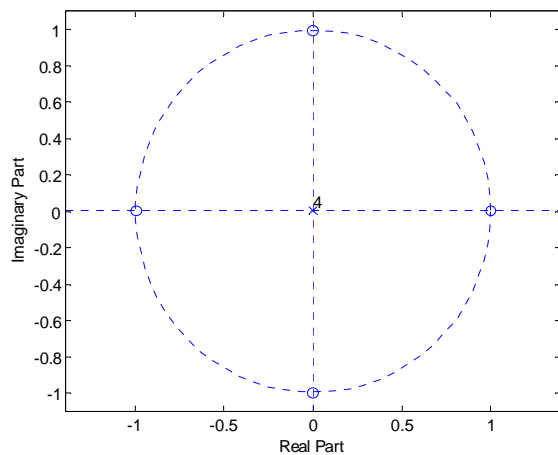
- Observe  $z=e^{j\omega}$  for  $-\pi < \omega < \pi$ :  $|z|=1$ , phase =  $\omega$
- This defines a *circle* in the z-plane with radius=1: referred to as the *unit circle*



# Visualizing Frequency Response

- We can observe z-transform along the unit circle to reveal the frequency response.

$$H(z) = 1 - z^{-4}$$



# Poles and Zeros

- A *pole* in the  $z$ -domain is a value of  $z$  that “pushes up” the magnitude like a tent pole.
- A *zero* in the  $z$ -domain is a value of  $z$  that “pins down” the magnitude like a stake or tack.
- The pole and zero locations control the magnitude everywhere, *including along the unit circle*.

# FIR Systems

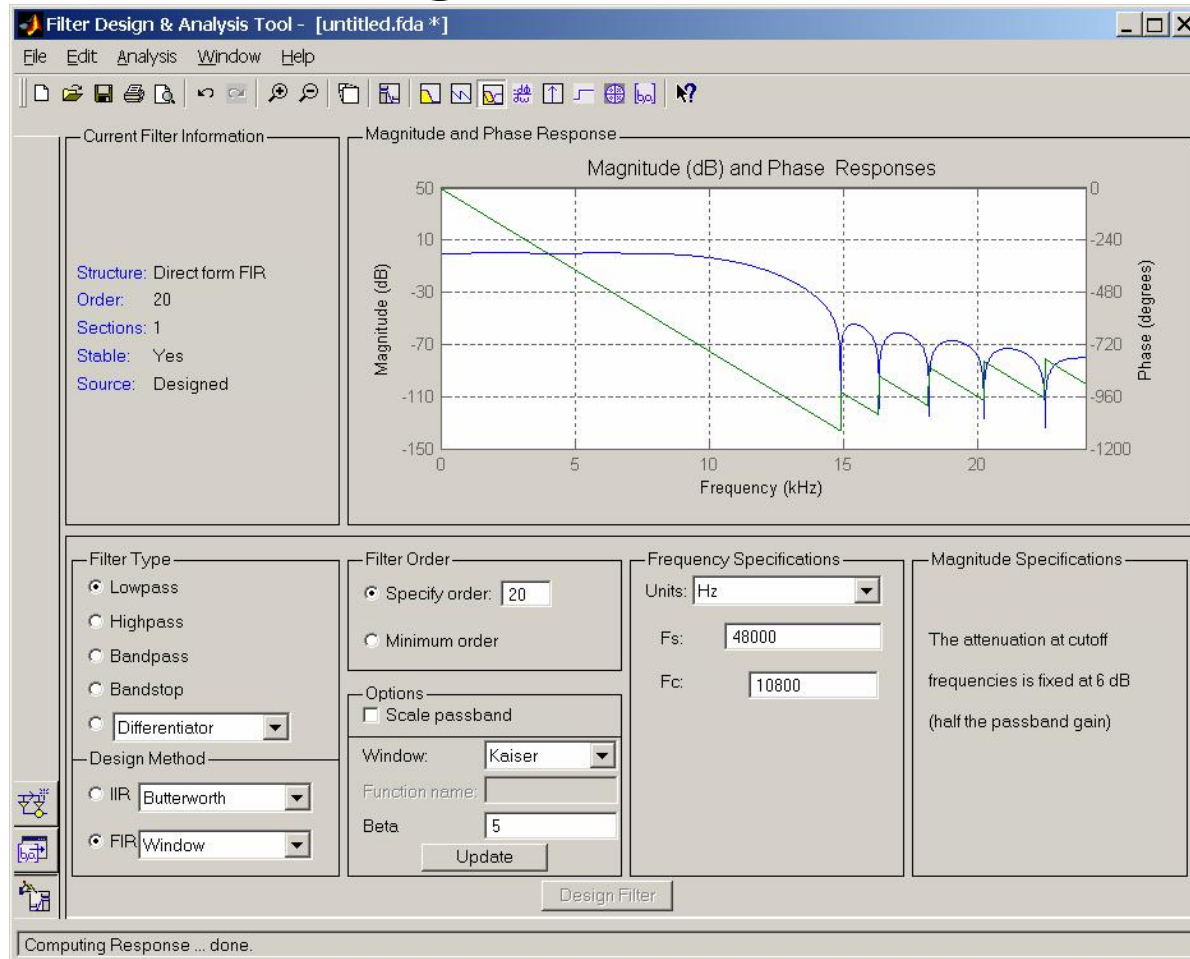
- FIR systems contain only finite zeros. Poles are located at zero (and perhaps infinity).
- FIR filter design requires a careful choice of zero locations.
- Stop band has zeros on the unit circle.
- Pass band has zeros off the unit circle.



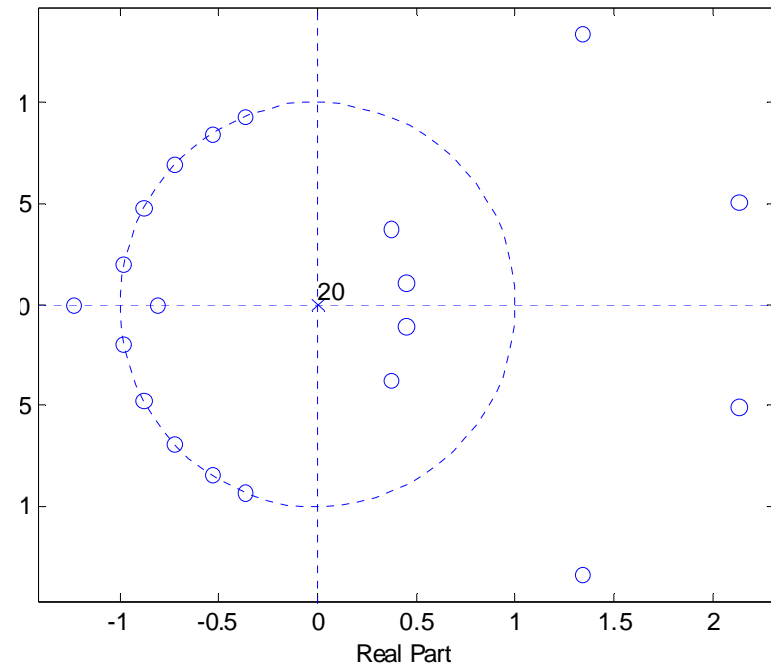
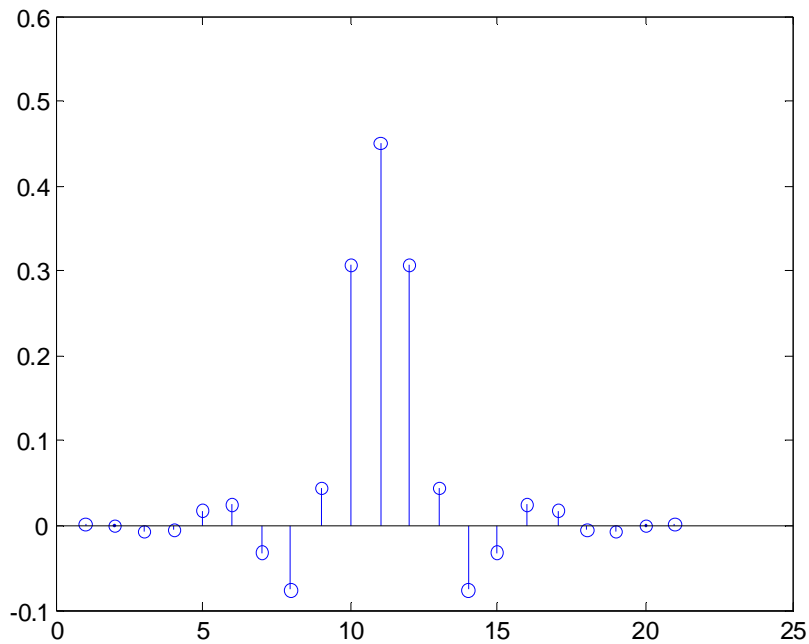
# Matlab FIR Filter Design

- Matlab provides several FIR filter design tools, including: `fir1`, `fir2`, and `remez`
- Matlab GUI: Filter Design and Analysis Tool (FDATool)
- Usually specify passband ripple, stopband attenuation, band edges, filter order, and  $f_s$

# Design Example



# Design Example (cont.)



# Symmetry and Linear Phase

- FIR systems with symmetric coefficients ( $b_k = b_{M-k}$ ) have frequency responses with *linear phase*.
- Show this by grouping z-transform terms, for example:

$$\begin{aligned} H(z) &= b_0 + b_1 z^{-1} + b_2 z^{-2} + b_3 z^{-3} + b_2 z^{-4} + b_1 z^{-5} + b_0 z^{-6} \\ &= z^{-3} [b_0 (z^3 + z^{-3}) + b_1 (z^2 + z^{-2}) + b_2 (z^1 + z^{-1}) + b_3] \end{aligned}$$

# Linear Phase (cont.)

- Now evaluate  $H(z)$  on unit circle:

$$H(e^{j\hat{\omega}}) = e^{-j3\hat{\omega}} \left[ b_0 (e^{j3\hat{\omega}} + e^{-j3\hat{\omega}}) + b_1 (e^{j2\hat{\omega}} + e^{-j2\hat{\omega}}) + b_2 (e^{j\hat{\omega}} + e^{-j\hat{\omega}}) + b_3 \right]$$

linear phase term
 $2\cos(3\hat{\omega})$ 
 $2\cos(2\hat{\omega})$ 
 $2\cos(\hat{\omega})$

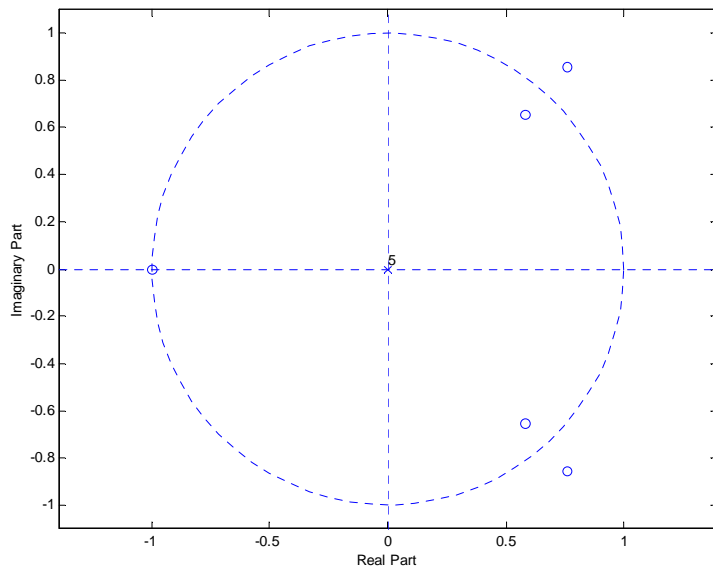
a real function of  $\hat{\omega}$  (phase=0)

- Example if  $M$  is odd:

$$\begin{aligned}
 H(z) &= b_0 + b_1 z^{-1} + b_2 z^{-2} + b_2 z^{-3} + b_1 z^{-4} + b_0 z^{-5} \\
 &= z^{-2.5} \left[ b_0 (z^{2.5} + z^{-2.5}) + b_1 (z^{1.5} + z^{-1.5}) + b_2 (z^{0.5} + z^{-0.5}) \right]
 \end{aligned}$$

# Zero Symmetry

- For an FIR linear phase system (implies coefficient symmetry), the zeros will have a specific pattern. For each  $z_0$ , there will be:



$$\left\{ z_0, z_0^*, \frac{1}{z_0}, \frac{1}{z_0^*} \right\}$$